FemtoClock®
Crystal/LVCMOS-to-LVPECL/ LVCMOS Frequency Synthesizer
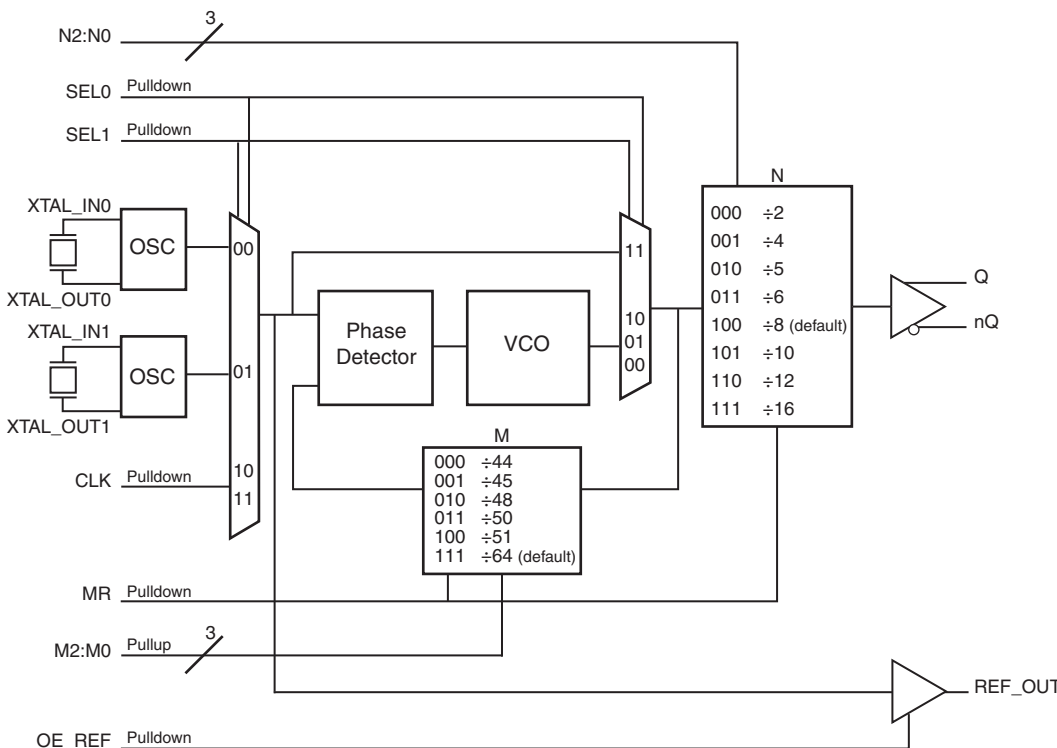
843001I-23

DATA SHEET

General Description

The 843001I-23 is a highly versatile, low phase noise LVPECL/LVCMOS Synthesizer which can generate low jitter reference clocks for a variety of communication applications. The dual crystal interface allows the synthesizer to support up to three communication standards in a given application (i.e. SONET with a 19.44MHz crystal, 1Gb/10Gb Ethernet and Fibre Channel using a 25MHz crystal). The RMS phase jitter performance is typically less than 1ps, thus making the device acceptable for use in demanding applications such as OC48 SONET, GbE/10Gb Ethernet and SAN applications. The 843001I-23 is packaged in a small 24-pin TSSOP, E-Pad package.

Features

- One 3.3Vdifferential LVPECL output pair and one LVCMOS/LVTTL single-ended reference clock output
- Selectable crystal oscillator interface or LVCMOS/LVTTL single-ended input
- Crystal and CLK range: 19.44MHz – 27MHz
- Able to generate GbE/10GbE/12GbE, Fibre Channel (1Gb/4Gb/10Gb), PCI-E and SATA from a 25MHz crystal
- VCO range: 1.12GHz – 1.275GHz
- Supports the following applications: SONET, Ethernet, Fibre Channel, Serial ATA, and HDTV
- RMS phase jitter @ 622.08MHz (12kHz - 20MHz): 0.9ps (typical), 3.3V
- Supply modes
  V_{CC}/V_{CCO}
  3.3V/3.3V
  3.3V/2.5V
  2.5V/2.5V
- -40°C to 85°C ambient operating temperature
- Available in lead-free (RoHS 6) package

Block Diagram

Pin Assignment

843001I-23

24-Lead TSSOP, E-Pad
4.4mm x 7.8mm x 0.925mm
package body
G Package
Top View

Table 1. Pin Descriptions

Number	Name	Type		Description
1	V_{CCO_LVCMOS}	Power		Output supply pin for REF_CLK output.
2, 3	N0, N1	Input	Pulldown	Output divider select pins. LVCMOS/LVTTL interface levels. See Table 3C.
4	N2	Input	Pullup	
5	V_{CCO_LVPECL}	Power		Output supply pin for LVPECL output.
6, 7	Q, nQ	Output		Differential output pair. LVPECL interface levels.
8, 23	V_{EE}	Power		Negative supply pins.
9	V_{CCA}	Power		Analog supply pin.
10	V_{CC}	Power		Core supply pin.
11, 12	XTAL_OUT1, XTAL_IN1	Input		Parallel resonant crystal interface. XTAL_OUT1 is the output, XTAL_IN1 is the input.
13, 14	XTAL_OUT0, XTAL_IN0	Input		Parallel resonant crystal interface. XTAL_OUT0 is the output, XTAL_IN0 is the input.
15	CLK	Input	Pulldown	Single-ended clock input. LVCMOS/LVTTL interface levels.
16, 17	SEL0, SEL1	Input	Pulldown	Input MUX select pins. LVCMOS/LVTTL interface levels. See Table 3D.
18	MR	Input	Pulldown	Active HIGH Master Reset. When logic HIGH, the internal dividers are reset causing the true output Q to go low and the inverted output nQ to go high. When logic LOW, the internal dividers and the outputs are enabled. LVCMOS/LVTTL interface levels.
19, 20, 21	M0, M1, M2	Input	Pullup	Feedback divider select pins. LVCMOS/LVTTL interface levels. See Table 3B.
22	OE_REF	Input	Pulldown	Reference clock output enable. Default LOW. See Table 3E. LVCMOS/LVTTL interface levels.
24	REF_OUT	Output		Reference clock output. LVCMOS/LVTTL interface levels.

NOTE: *Pullup and Pulldown* refer to internal input resistors. See Table 2, *Pin Characteristics,* for typical values.

Table 2. Pin Characteristics

Symbol	Parameter	Test Conditions		Minimum	Typical	Maximum	Units
C_{IN}	Input Capacitance				4		pF
R_{PULLUP}	Input Pullup Resistor				51		kΩ
$R_{PULLDOWN}$	Input Pulldown Resistor				51		kΩ
R_{OUT}	Output Impedance	REF_OUT	V_{CCO} = 3.3V		21		Ω
			V_{CCO} = 2.5V		25		Ω

Function Tables

Table 3A. Common Configuration Table

Input Frequency (MHz)	M Feedback Divider Value	VCO Frequency (MHz)	N Output Divider Value	Output Frequency (MHz)	Application
27	44	1188	16	74.25	HDTV
24.75	48	1188	16	74.25	HDTV
19.44	64	1244.16	8	155.52	SONET
19.44	64	1244.16	2	622.08	SONET
19.44	64	1244.16	4	311.04	SONET
25	50	1250	10	125	GigE
25	50	1250	8	156.25	10 GigE
25	50	1250	5	250	GigE
25	50	1250	4	312.5	XGMII
25	50	1250	2	625	10 GigE
25	45	1125	6	187.5	12 GigE
25	48	1200	12	100	PCI Express
25	48	1200	8	150	SATA
25	48	1200	16	75	SATA
25	51	1275	12	106.25	Fibre Channel
25	51	1275	8	159.375	10 Gig Fibre Channel
25	51	1275	6	212.5	4 Gig Fibre Channel

Table 3B. Programmable M Feedback Divider Function Table

Inputs			M Feedback Divider Value	Input Frequency (MHz)	
M2	M1	M0		Minimum	Maximum
0	0	0	44	25.5	27
0	0	1	45	24.9	27
0	1	0	48	23.3	26.56
0	1	1	50	22.4	25.5
1	0	0	51	22.0	25
1	0	1	64 (default)	19.44	19.92

Table 3C. Programmable N Output Divider Function Table

Inputs			
N2	**N1**	**N0**	**N Divider Value**
0	0	0	2
0	0	1	4
0	1	0	5
0	1	1	6
1	0	0	8 (default)
1	0	1	10
1	1	0	12
1	1	1	16

Table 3D. Select Mode Function Table

Inputs			
SEL1	**SEL0**	**Reference Input**	**PLL Mode**
0	0	XTAL0	Active (default)
0	1	XTAL1	Active
1	0	CLK	Active
1	1	CLK	Bypass

Table 3E. OE_REF Output Function Table

Input	Output
OE_REF	**REF_OUT**
0	High-Impedance (default)
1	Active

 FEMTOCLOCK® CRYSTAL/LVCMOS-TO-LVPECL/ LVCMOS
FREQUENCY SYNTHESIZER

Absolute Maximum Ratings

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device.
These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond
those listed in the *DC Characteristics or AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for
extended periods may affect product reliability.

Item	Rating
Supply Voltage, V_{CC}	4.6V
Inputs, V_I XTAL_IN Other Input	 0V to V_{CC} -0.5V to V_{CC} + 0.5V
Outputs, I_O (LVPECL) Continuous Current Surge Current Outputs, V_O (LVCMOS)	 50mA 100mA -0.5V to V_{CCO_LVCMOS} + 0.5V
Package Thermal Impedance, θ_{JA}	32.1°C/W (0 mps)
Storage Temperature, T_{STG}	-65°C to 150°C

DC Electrical Characteristics

Table 4A. Power Supply DC Characteristics, $V_{CC} = V_{CCO\_LVCMOS} = V_{CCO\_LVPECL} = 3.3V \pm 5\%$, $V_{EE} = 0V$,
$T_A = -40°C$ to $85°C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{CC}	Core Supply Voltage		3.135	3.3	3.465	V
V_{CCA}	Analog Supply Voltage		V_{CC} − 0.11	3.3	V_{CC}	V
V_{CCO_LVPECL}, V_{CCO_LVCMOS}	Output Supply Voltage		3.135	3.3	3.465	V
I_{EE}	Power Supply Current				140	mA
I_{CCA}	Analog Supply Current	Outputs Unterminated			11	mA

Table 4B. Power Supply DC Characteristics, $V_{CC} = 3.3V \pm 5\%$, $V_{CCO\_LVCMOS} = V_{CCO\_LVPECL} = 2.5V \pm 5\%$, $V_{EE} = 0V$,
$T_A = -40°C$ to $85°C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{CC}	Core Supply Voltage		3.135	3.3	3.465	V
V_{CCA}	Analog Supply Voltage		V_{CC} − 0.11	3.3	V_{CC}	V
V_{CCO_LVPECL}, V_{CCO_LVCMOS}	Output Supply Voltage		2.375	2.5	2.625	V
I_{EE}	Power Supply Current				139	mA
I_{CCA}	Analog Supply Current	Outputs Unterminated			11	mA

Table 4C. Power Supply DC Characteristics, $V_{CC} = V_{CCO\_LVCMOS} = V_{CCO\_LVPECL} = 2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40°C$ to $85°C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{CC}	Core Supply Voltage		2.375	2.5	2.625	V
V_{CCA}	Analog Supply Voltage		$V_{CC} - 0.10$	2.5	V_{CC}	V
V_{CCO_PECL}, V_{CCO_CMOS}	Output Supply Voltage		2.375	2.5	2.625	V
I_{EE}	Power Supply Current				133	mA
I_{CCA}	Analog Supply Current	Outputs Unterminated			10	mA

Table 4D. LVCMOS/LVTTL DC Characteristics, $T_A = -40°C$ to $85°C$

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V_{IH}	Input High Voltage		$V_{CC} = 3.3V$	2		$V_{CC} + 0.3$	V
			$V_{CC} = 2.5V$	1.7		$V_{CC} + 0.3$	V
V_{IL}	Input Low Voltage		$V_{CC} = 3.3V$	-0.3		0.8	V
			$V_{CC} = 2.5V$	-0.3		0.7	V
I_{IH}	Input High Current	CLK, OE_REF, MR, N0, N1 SEL0, SEL1	$V_{CC} = V_{IN} = 3.465V$ or $2.625V$			150	µA
		N2, M[2:0]	$V_{CC} = V_{IN} = 3.465V$ or $2.625V$			5	µA
I_{IL}	Input Low Current	CLK, OE_REF, MR, N0, N1 SEL0, SEL1	$V_{CC} = 3.465V$ or $2.625V$, $V_{IN} = 0V$	-5			µA
		N2, M[2:0]	$V_{CC} = 3.465V$ or $2.625V$, $V_{IN} = 0V$	-150			µA
V_{OH}	Output High Voltage	REF_OUT	$V_{CCO\_LVCMOS} = 3.465V$, $I_{OH} = -12mA$	2.6			V
			$V_{CCO\_LVCMOS} = 2.625V$, $I_{OH} = -12mA$	1.8			V
V_{OL}	Output Low Voltage	REF_OUT	$V_{CCO\_LVCMOS} = 3.465V$ or $2.625V$, $I_{OL} = 12mA$			0.5	V

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Table 4E. LVPECL DC Characteristics, $V_{CC} = V_{CCO\_LVPECL} = 3.3V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40°C$ to $85°C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{OH}	Output High Voltage; NOTE 1		$V_{CCO\_LVPECL} - 1.4$		$V_{CCO\_LVPECL} - 0.9$	V
V_{OL}	Output Low Voltage; NOTE 1		$V_{CCO\_LVPECL} - 2.0$		$V_{CCO\_LVPECL} - 1.7$	V
V_{SWING}	Peak-to-Peak Output Voltage Swing		0.6		1.0	V

NOTE 1: Outputs terminated with 50Ω to $V_{CCO\_LVPECL} - 2V$.

Table 4F. LVPECL DC Characteristics, V_{CC} = 3.3V ± 5% or 2.5V ± 5%, V_{CCO_LVPECL} = 2.5V ± 5%, V_{EE} = 0V, T_A = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{OH}	Output High Voltage; NOTE 1		$V_{CCO\_LVPECL} - 1.4$		$V_{CCO\_LVPECL} - 0.9$	V
V_{OL}	Output Low Voltage; NOTE 1		$V_{CCO\_LVPECL} - 2.0$		$V_{CCO\_LVPECL} - 1.5$	V
V_{SWING}	Peak-to-Peak Output Voltage Swing		0.4		1.0	V

NOTE 1: Outputs terminated with 50Ω to $V_{CCO\_LVPECL} - 2V$.

Table 5. Crystal Characteristics

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation			Fundamental		
Frequency		19.44		27	MHz
Equivalent Series Resistance (ESR)				50	Ω
Shunt Capacitance				7	pF

NOTE: Characterized using an 18pF parallel resonant crystal.

AC Electrical Characteristics

Table 6A. AC Characteristics, V_{CC} = V_{CCO_LVCMOS} = V_{CCO_LVPECL} = 3.3V ± 5%, V_{EE} = 0V, T_A = -40°C to 85°C

| Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
f_{OUT}	Output Frequency	Q, nQ		70		637.5	MHz
		REF_OUT		19.44		27	MHz
t_{PD}	Propagation Delay; NOTE 1	CLK to REF_OUT		2.2		2.7	ns
$f_{jit}(\emptyset)$	RMS Phase Jitter, (Random); NOTE 2		622.08MHz, (12kHz – 20MHz)		0.97		ps
f_{VCO}	PLL VCO Lock Range			1.12		1.275	GHz
t_R / t_F	Output Rise/Fall Time	Q, nQ	20% to 80%	200		700	ps
		REF_OUT, NOTE 3	20% to 80%	250		650	ps
odc	Output Duty Cycle	Q, nQ		46		54	%
		REF_OUT; NOTE 3	Using Clock Input	48		52	%
t_{LOCK}	PLL Lock Time					60	ms

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.
NOTE 1: Measured from the V_{CC}/2 of the input to V_{CCO_LVCMOS}/2 of the output.
NOTE 2: Phase jitter measured using a 19.44MHz quartz crystal.
NOTE 3: REF_OUT output duty cycle characterized with CLK input duty cycle between 48% and 52%.

**Table 6B. AC Characteristics,** V_{CC} = 3.3V ± 5%, V_{CCO_LVCMOS} = V_{CCO_LVPECL} = 2.5V ± 5%, V_{EE} = 0V, T_A = -40°C to 85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
f_{OUT}	Output Frequency	Q, nQ		70		637.5	MHz
		REF_OUT		19.44		27	MHz
t_{PD}	Propagation Delay; NOTE 1	CLK to REF_OUT		2.3		2.9	ns
fjit(Ø)	RMS Phase Jitter, (Random); NOTE 2		622.08MHz, (12kHz – 20MHz)		1		ps
f_{VCO}	PLL VCO Lock Range			1.12		1.275	GHz
t_R / t_F	Output Rise/Fall Time	Q, nQ	20% to 80%	200		700	ps
		REF_OUT	20% to 80%	350		750	ps
odc	Output Duty Cycle	Q, nQ		46		54	%
		REF_OUT; NOTE 3	Using Clock Input	48		52	%
t_{LOCK}	PLL Lock Time					60	ms

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.
NOTE 1: Measured from the V_{CC}/2 of the input to V_{CCO_LVCMOS}/2 of the output.
NOTE 2: Phase jitter measured using a 19.44MHz quartz crystal.
NOTE 3: REF_OUT output duty cycle characterized with CLK input duty cycle between 48% and 52%.

Table 6C. AC Characteristics, V_{CC} = V_{CCO_LVCMOS} = V_{CCO_LVPECL} = 2.5V ± 5%, V_{EE} = 0V, T_A = -40°C to 85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
f_{OUT}	Output Frequency	Q, nQ		70		637.5	MHz
		REF_OUT		19.44		27	MHz
t_{PD}	Propagation Delay; NOTE 1	CLK to REF_OUT		2.3		2.9	ns
fjit(Ø)	RMS Phase Jitter, (Random); NOTE 2		622.08MHz, (12kHz – 20MHz)		1.1		ps
f_{VCO}	PLL VCO Lock Range			1.12		1.275	GHz
t_R / t_F	Output Rise/Fall Time	Q, nQ	20% to 80%	200		700	ps
		REF_OUT	20% to 80%	350		750	ps
odc	Output Duty Cycle	Q, nQ		46		54	%
		REF_OUT; NOTE 3	Using Clock Input	48		52	%
t_{LOCK}	PLL Lock Time					60	ms

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.
NOTE 1: Measured from the V_{CC}/2 of the input to V_{CCO_LVCMOS}/2 of the output.
NOTE 2: Phase jitter measured using a 19.44MHz quartz crystal.
NOTE 3: REF_OUT output duty cycle characterized with CLK input duty cycle between 48% and 52%.

Typical Phase Noise at 622.08MHz

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622.08MHz
RMS Phase Jitter (Random)
12kHz to 20MHz = 0.97ps (typical)

Offset Frequency (Hz)
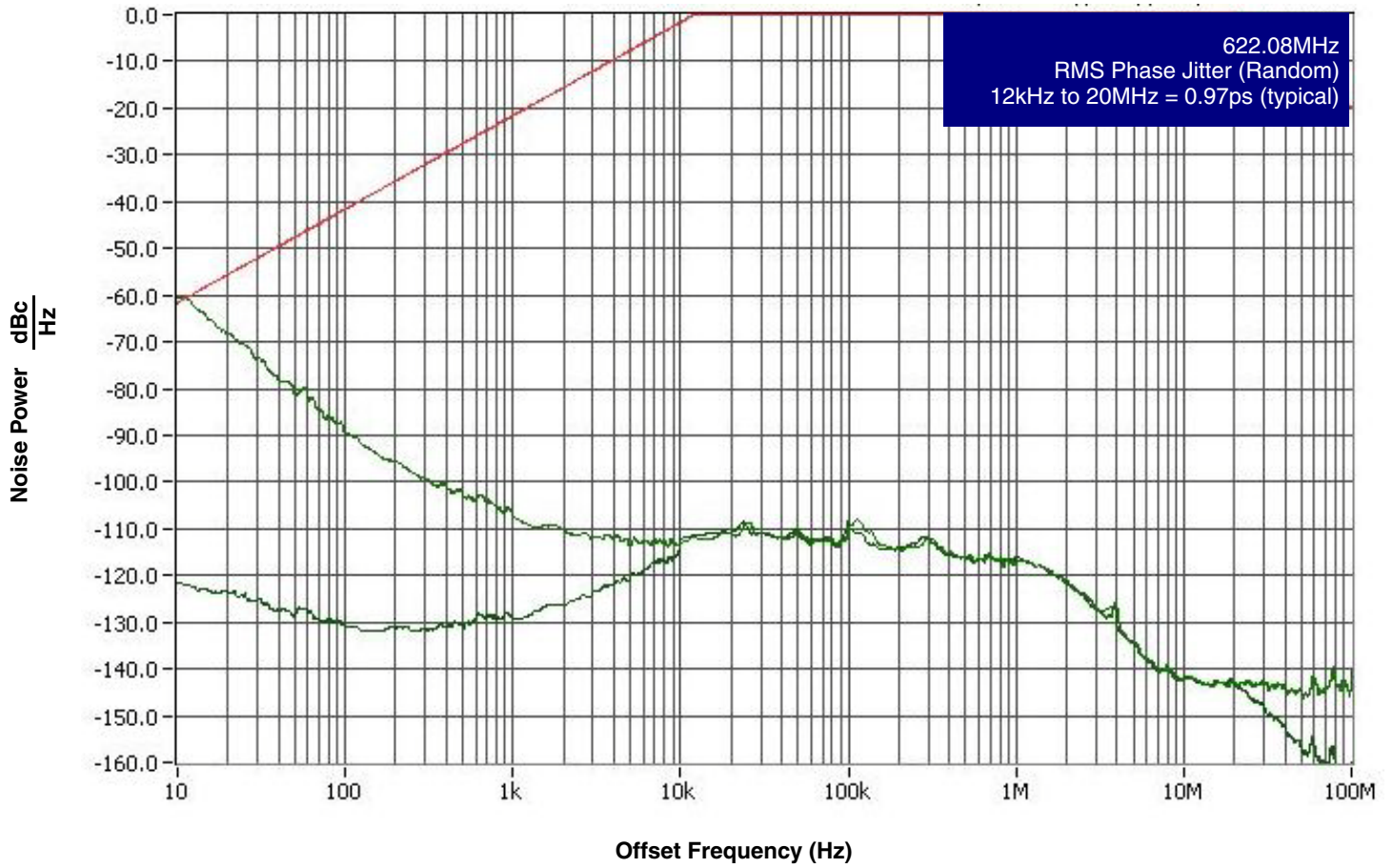
Parameter Measurement Information
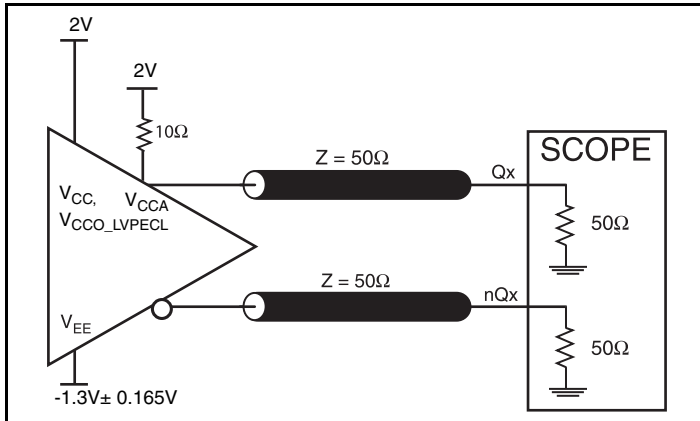
3.3V LVPECL Output Load AC Test Circuit
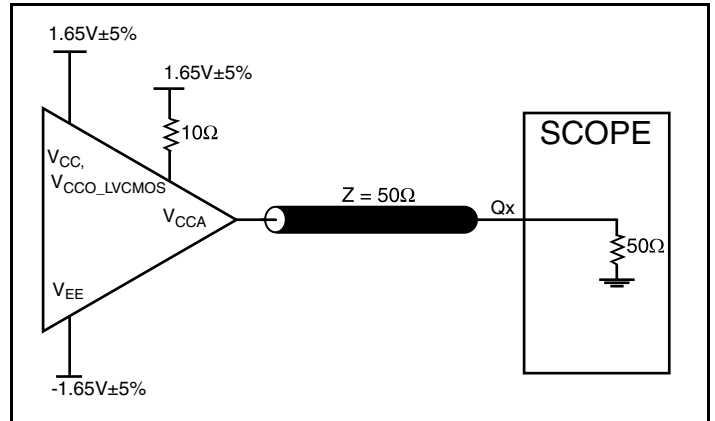
3.3V LVCMOS Output Load AC Test Circuit
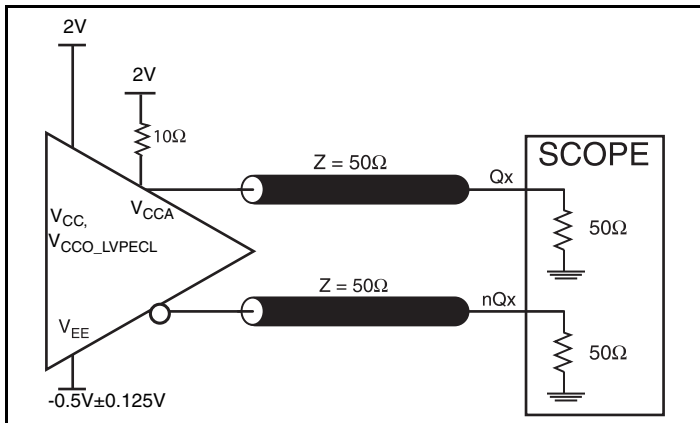
2.5V LVPECL Output Load AC Test Circuit
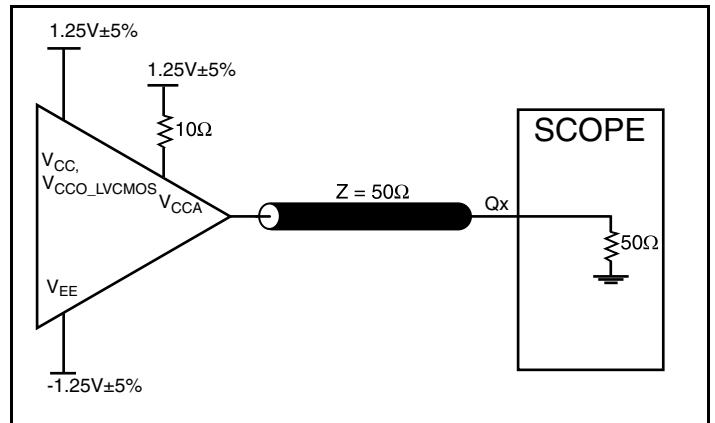
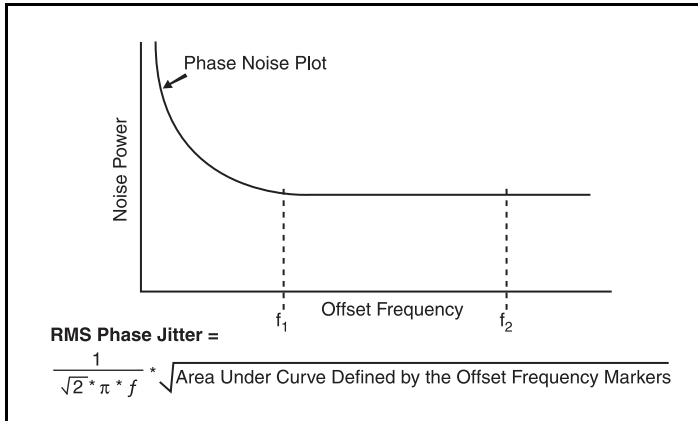

2.5V LVCMOS Output Load AC Test Circuit
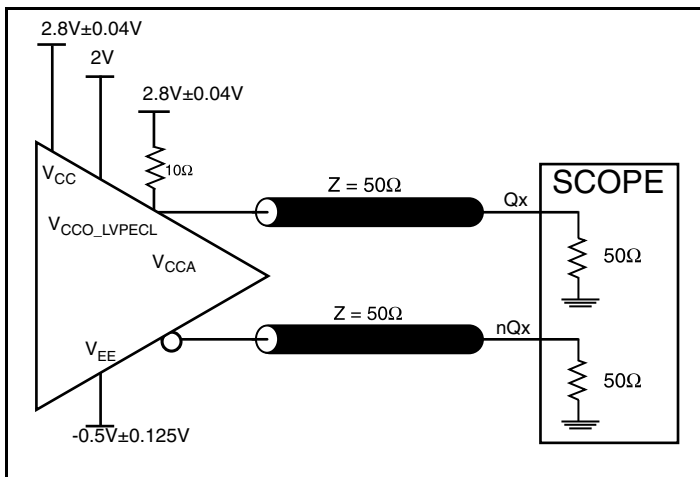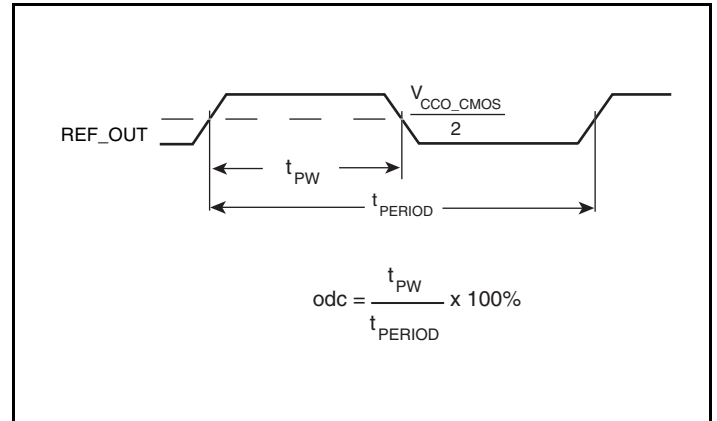

3.3 Core/2.5V LVPECL Output Load AC Test Circuit
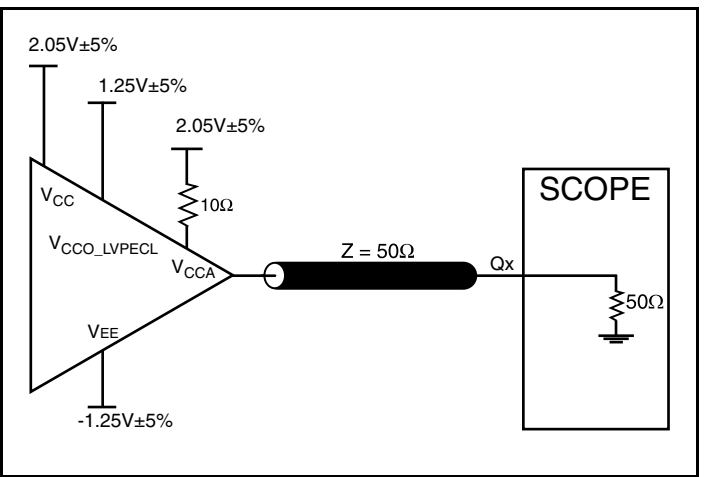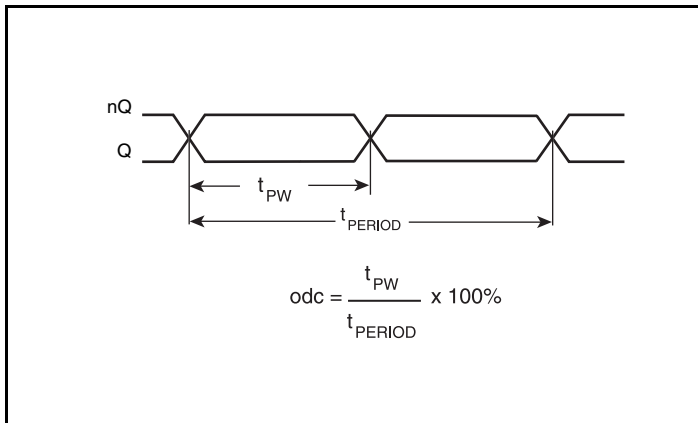

3.3V Core/2.5V LVCMOS Output Load AC Test Circuit

Parameter Measurement Information, continued

RMS Phase Jitter

LVCMOS Output Duty Cycle/Pulse Width/Period

LVPECL Output Duty Cycle/Pulse Width/Period
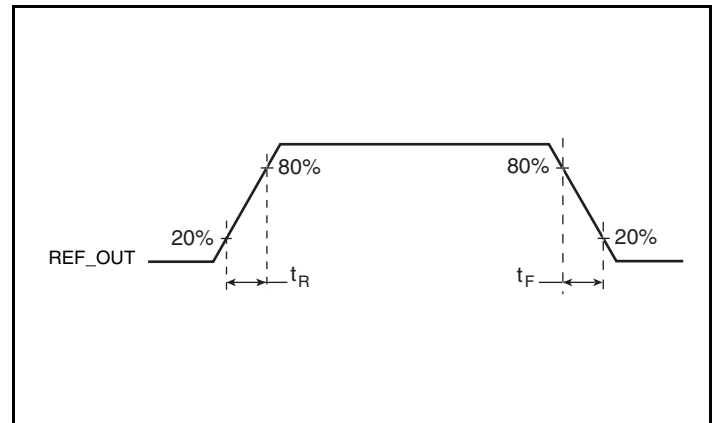
LVCMOS Output Rise/Fall Time
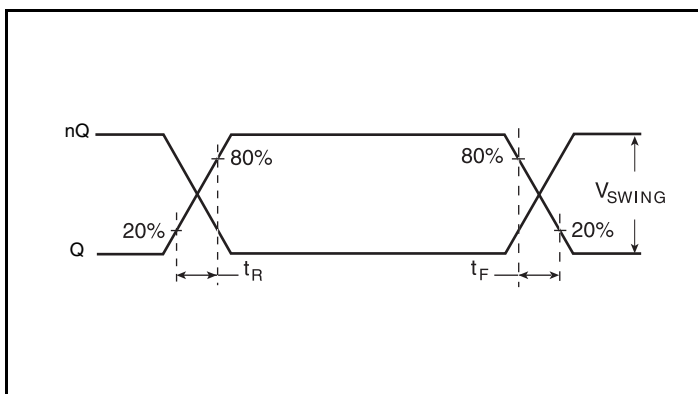
LVPECL Output Rise/Fall Time

Applications Information

Recommendations for Unused Input and Output Pins

Inputs:

Crystal Inputs

For applications not requiring the use of the crystal oscillator input, both XTAL_IN and XTAL_OUT can be left floating. Though not required, but for additional protection, a 1kΩ resistor can be tied from XTAL_IN to ground.

CLK Input

For applications not requiring the use of the clock input, it can be left floating. Though not required, but for additional protection, a 1kΩ resistor can be tied from the CLK input to ground.

LVCMOS Control Pins

All control pins have internal pullups or pulldowns; additional resistance is not required but can be added for additional protection. A 1kΩ resistor can be used.

Outputs:

LVPECL Outputs

The unused LVPECL output pair can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

LVCMOS Output

All unused LVCMOS output can be left floating. We recommend that there is no trace attached.

Overdriving the XTAL Interface

The XTAL_IN input can accept a single-ended LVCMOS signal through an AC coupling capacitor. A general interface diagram is shown in *Figure 1A*. The XTAL_OUT pin can be left floating. The maximum amplitude of the input signal should not exceed 2V and the input edge rate can be as slow as 10ns. This configuration requires that the output impedance of the driver (Ro) plus the series resistance (Rs) equals the transmission line impedance. In addition,

matched termination at the crystal input will attenuate the signal in half. This can be done in one of two ways. First, R1 and R2 in parallel should equal the transmission line impedance. For most 50Ω applications, R1 and R2 can be 100Ω. This can also be accomplished by removing R1 and making R2 50Ω. By overdriving the crystal oscillator, the device will be functional, but note, the device performance is guaranteed by using a quartz crystal.
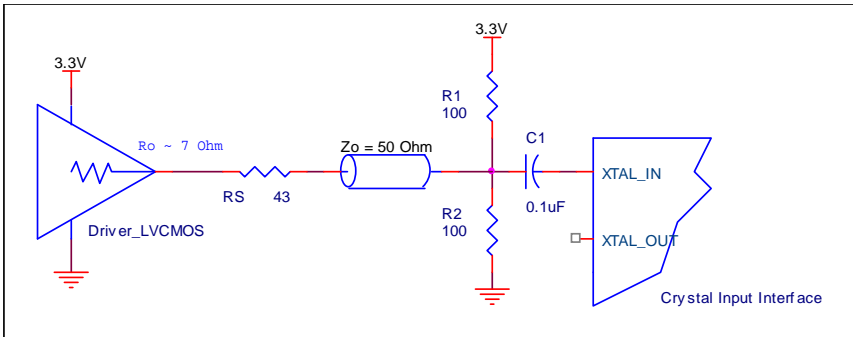
Figure 1A. General Diagram for LVCMOS Driver to XTAL Input Interface
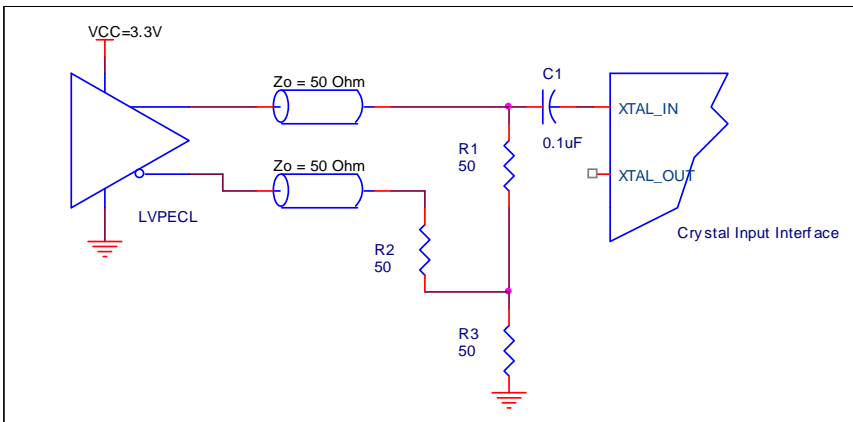
Figure 1B. General Diagram for LVPECL Driver to XTAL Input Interface

Termination for 3.3V LVPECL Outputs

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

The differential outputs are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive 50Ω

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 2A and 2B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.
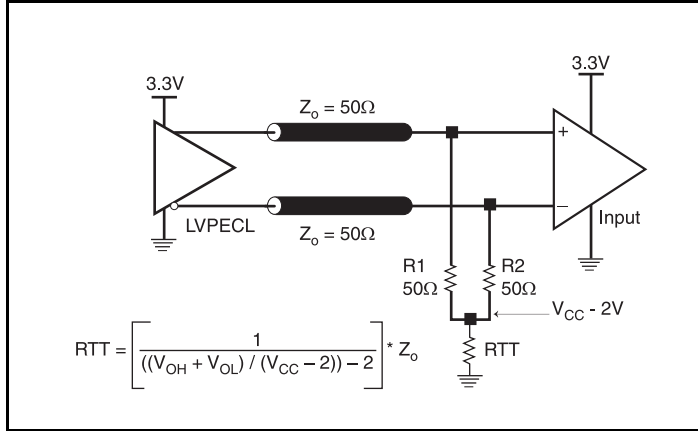
Figure 2A. 3.3V LVPECL Output Termination

$$RTT = \left[\frac{1}{((V_{OH} + V_{OL}) / (V_{CC} - 2)) - 2} \right] * Z_o$$
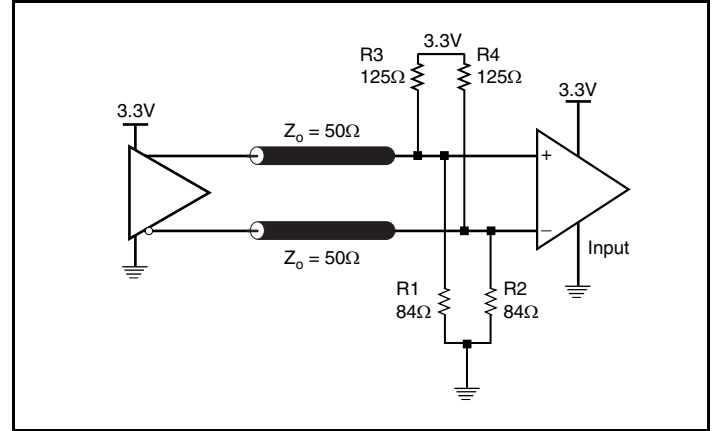
Figure 2B. 3.3V LVPECL Output Termination

Termination for 2.5V LVPECL Outputs

Figure 3A and *Figure 3B* show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating 50Ω to $V_{CCO} - 2V$. For $V_{CCO} = 2.5V$, the $V_{CCO} - 2V$ is very close to ground level. The R3 in Figure 3B can be eliminated and the termination is shown in *Figure 3C.*
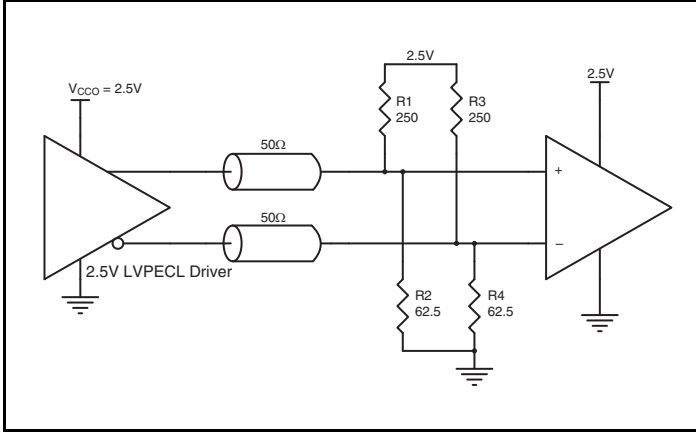
Figure 3A. 2.5V LVPECL Driver Termination Example
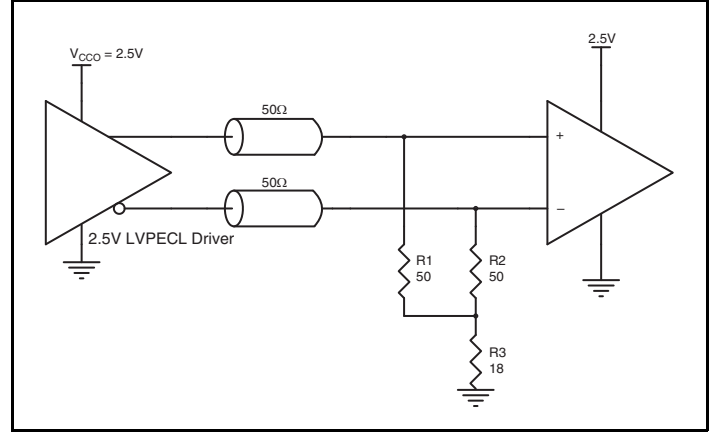
Figure 3B. 2.5V LVPECL Driver Termination Example
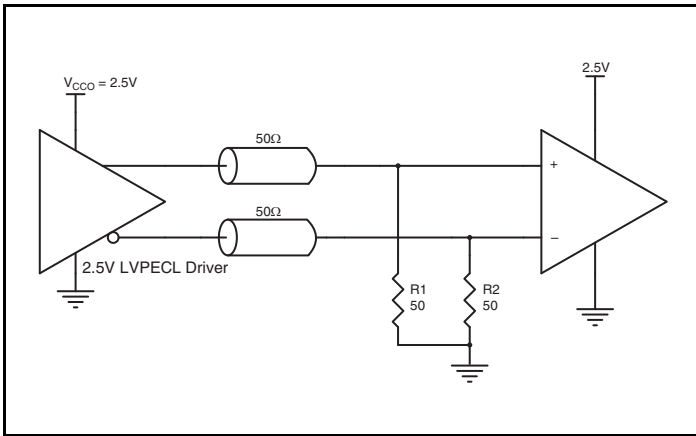
Figure 3C. 2.5V LVPECL Driver Termination Example

Schematic Layout

Figure 6 (next page) shows an example of 843001I-23 application schematic. In this example, the device is operated V_{CC} = V_{CCO_LVCMOS} = V_{CCO_LVPECL} = 3.3V. The 18pF parallel resonant 17.5-29.54MHz crystal is used. The load capacitance C1 = 22pF and C2 = 22pF are recommended for frequency accuracy. Depending on the parasitic of the printed circuit board layout, these values might require a slight adjustment to optimize the frequency accuracy. Crystals with other load capacitance specifications can be used. This will require adjusting C1 and C2. For this device, the crystal load capacitors are required for proper operation.

As with any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. The 843001I-23 provides separate power supplies to isolate any high switching noise from coupling into the internal PLL.

In order to achieve the best possible filtering, it is recommended that the placement of the filter components be on the device side of the PCB as close to the power pins as possible. If space is limited, the 0.1uF capacitor in each power pin filter should be placed on the device side. The other components can be on the opposite side of the PCB.

Power supply filter recommendations are a general guideline to be used for reducing external noise from coupling into the devices. The filter performance is designed for wide range of noise frequency. This low-pass filter starts to attenuate noise at approximately 10kHz. If a specific frequency noise component with high amplitude interference is known, such as switching power supplies frequencies, it is recommended that component values be adjusted and if required, additional filtering be added. Additionally general design practice for power plane voltage stability suggests adding bulk capacitances in the general area of all devices.

The schematic example focuses on functional connections and is not configuration specific. Refer to the pin description and functional tables in the datasheet to ensure the logic control inputs are properly set.
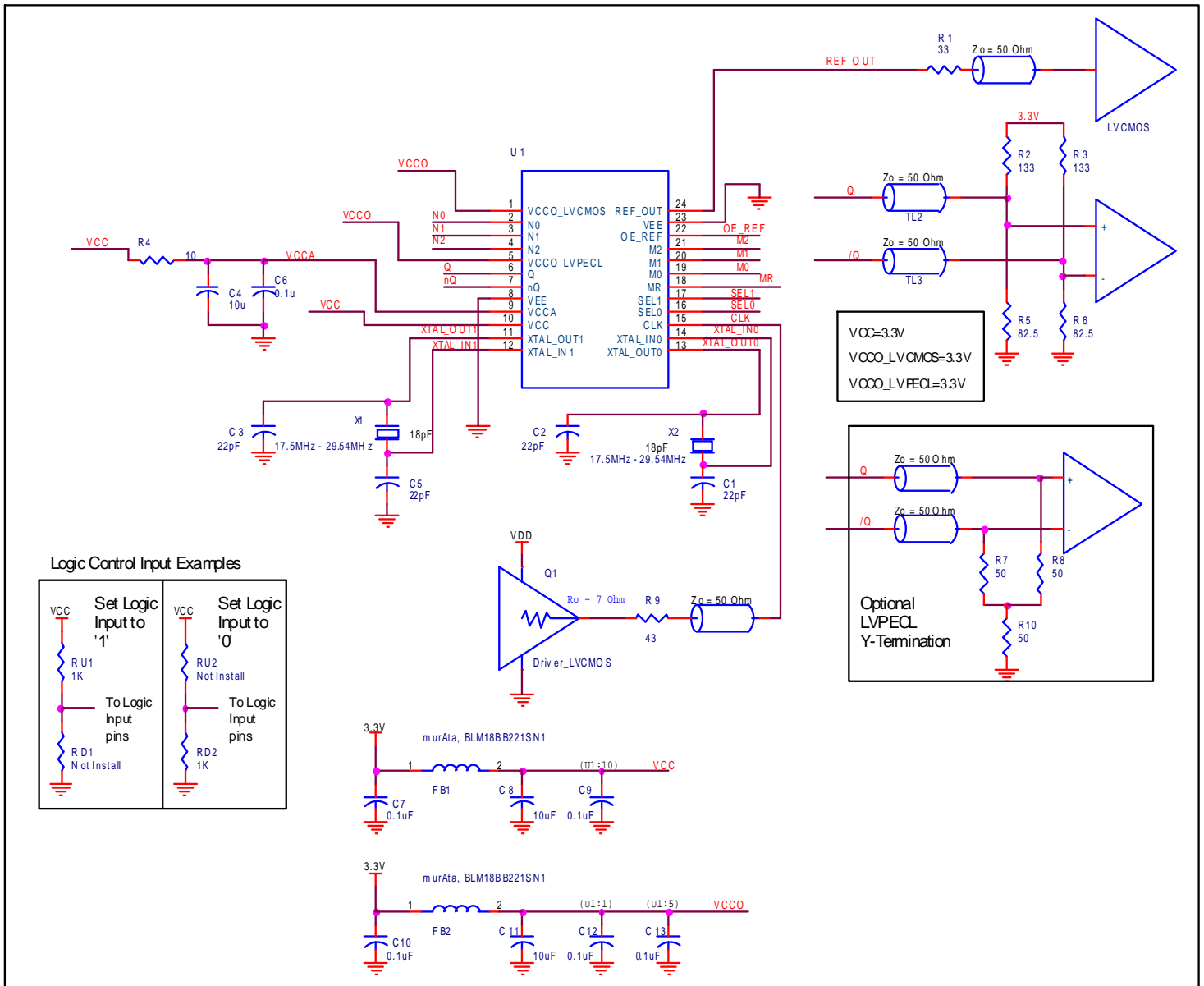
Figure 6. 843001I-23 Layout Example

EPAD Thermal Release Path

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in *Figure 7.* The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The vias act as "heat pipes". The number of vias (i.e. "heat pipes") are application specific and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, refer to the Application Note on the *Surface Mount Assembly* of Amkor's Thermally/Electrically Enhance Leadframe Base Package, Amkor Technology.
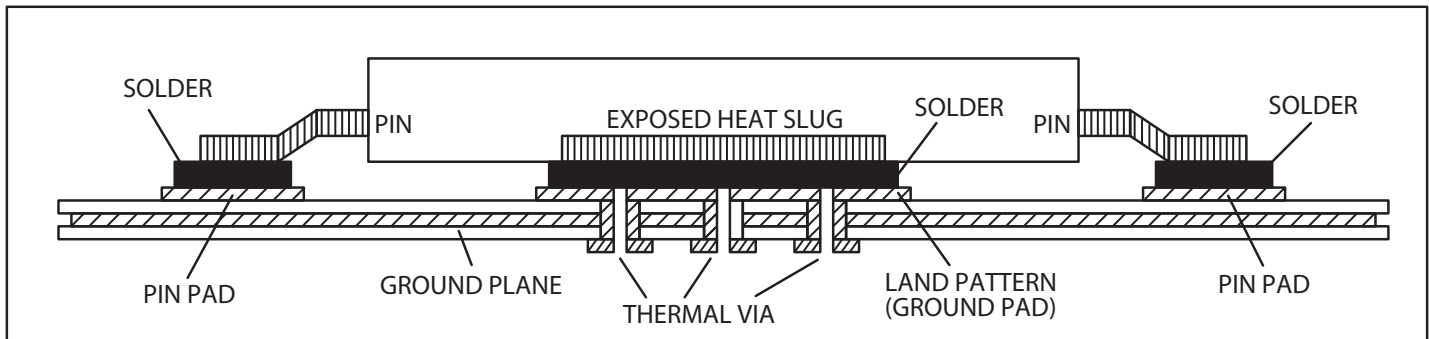
Figure 7. Assembly for Exposed Pad Thermal Release Path - Side View (drawing not to scale)

Power Considerations

This section provides information on power dissipation and junction temperature for the 843001I-23.
Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the 843001I-23 is the sum of the core power plus the power dissipated in the load(s).
The following is the power dissipation for V_{CC} = 3.3V + 5% = 3.465V, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)$_{MAX}$ = V_{CC_MAX} * I_{EE_MAX} = 3.465V * 140mA = **485.1mW**

- Power (outputs)$_{MAX}$ = **30mW/Loaded Output pair**

LVCMOS Output Power Dissipation

- Output Impedance R_{OUT} Power Dissipation due to Loading 50Ω to V_{DDO}/2
  Output Current I_{OUT} = V_{DDO_MAX} / [2 * (50Ω + R_{OUT})] = 3.465V / [2 * (50Ω + 21Ω)] = **24.4mA**

- Power Dissipation on the R_{OUT} per LVCMOS output
  Power (R_{OUT}) = R_{OUT} * (I_{OUT})^2 = 21Ω * (24.4mA)^2 = **12.5mW per output**

Total Power Dissipation

- **Total Power**
  = Power (core) + Power (LVPECL output) + Power (R_{OUT})
  = 485.1mW + 30mW + 12.5mW **= 527.6mW**

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad, and directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj = θ_{JA} * Pd_total + T$_A$

Tj = Junction Temperature

θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

T$_A$ = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 32.1°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

85°C + 0.528W * 32.1°C/W = 102°C. This is below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 7. Thermal Resitance θ_{JA} for 24 Lead TSSOP, E-Pad Forced Convection

θ_{JA} **vs. Air Flow**			
Meters per Second	**0**	**1**	**2.5**
Multi-Layer PCB, JEDEC Standard Test Boards	32.1°C/W	25.5°C/W	24.0°C/W

3. Calculations and Equations.

The purpose of this section is to calculate the power dissipation for the LVPECL output pair.

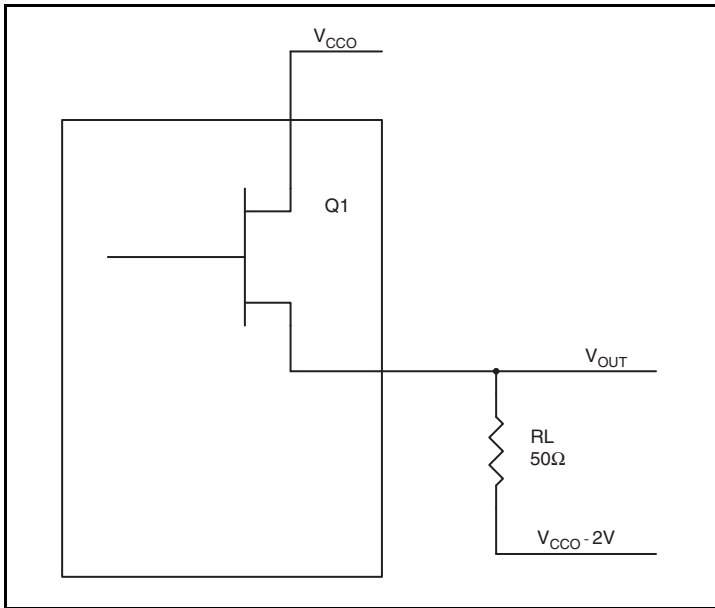LVPECL output driver circuit and termination are shown in *Figure 7.*

Figure 8. LVPECL Driver Circuit and Termination

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of $V_{CCO} - 2V$.

- For logic high, $V_{OUT} = V_{OH\_MAX} = \textbf{V}_{\textbf{CCO\_MAX}} - \textbf{0.9V}$
  $(V_{CCO\_MAX} - V_{OH\_MAX}) = \textbf{0.9V}$

- For logic low, $V_{OUT} = V_{OL\_MAX} = \textbf{V}_{\textbf{COO\_MAX}} - \textbf{1.7V}$
  $(V_{CCO\_MAX} - V_{OL\_MAX}) = \textbf{1.7V}$

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

$Pd\_H = [(V_{OH\_MAX} - (V_{CCO\_MAX} - 2V))/R_L] * (V_{CCO\_MAX} - V_{OH\_MAX}) = [(2V - (V_{CCO\_MAX} - V_{OH\_MAX}))/R_L] * (V_{CCO\_MAX} - V_{OH\_MAX}) = [(2V - 0.9V)/50\Omega] * 0.9V = \textbf{19.8mW}$

$Pd\_L = [(V_{OL\_MAX} - (V_{CCO\_MAX} - 2V))/R_L] * (V_{CCO\_MAX} - V_{OL\_MAX}) = [(2V - (V_{CCO\_MAX} - V_{OL\_MAX}))/R_L] * (V_{CCO\_MAX} - V_{OL\_MAX}) = [(2V - 1.7V)/50\Omega] * 1.7V = \textbf{10.2mW}$

Total Power Dissipation per output pair = Pd_H + Pd_L = **30mW**

Reliability Information

Table 8. θ_{JA} vs. Air Flow Table for a 24 Lead TSSOP, E-pad

θ_{JA} vs. Air Flow			
Meters per Second	**0**	**1**	**2.5**
Multi-Layer PCB, JEDEC Standard Test Boards	32.1°C/W	25.5°C/W	24.0°C/W

Transistor Count

The transistor count for 843001I-23 is: 4165

Package Outline and Package Dimensions

Package Outline - G Suffix for 24 Lead TSSOP, E-Pad

TOP VIEW

SIDE VIEW

EXPOSED PAD VIEW

Table 9. Package Dimensions

All Dimensions in Millimeters		
Symbol	Minimum	Maximum
N	24	
A		1.10
A1	0.05	0.15
A2	0.85	0.95
b	0.19	0.30
b1	0.19	0.25
c	0.09	0.20
c1	0.09	0.16
D	7.70	7.90
E	6.40 Basic	
E1	4.30	4.50
e	0.65 Basic	
L	0.50	0.70
P	5.0	5.5
P1	3.0	3.2
α	0°	8°
aaa	0.076	
bbb	0.10	

Reference Document: JEDEC Publication 95, MO-153

EVEN LEAD SIDES
TOPVIEW

ODD LEAD SIDES
TOPVIEW

SECTION "B-B"
SCALE: 120/1
(SEE NOTE 10)

DETAIL 'A'
(SCALE: 30/1)
(VIEW ROTATED 90° C.W.)

END VIEW
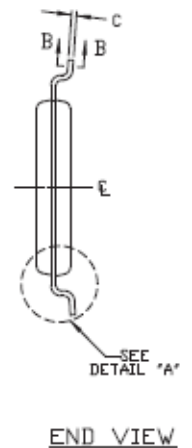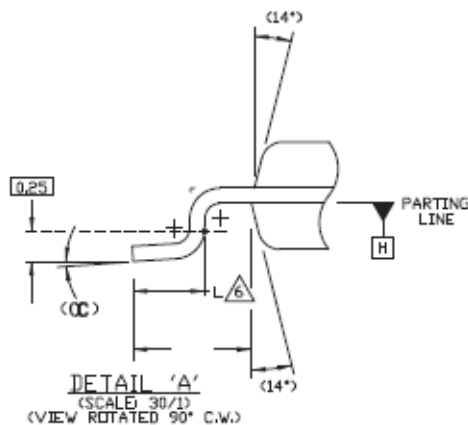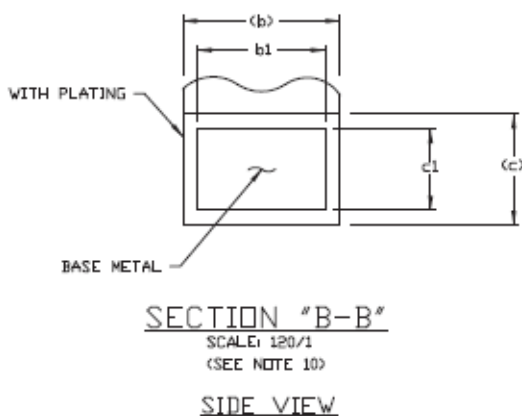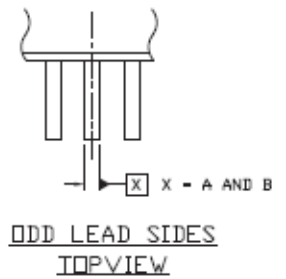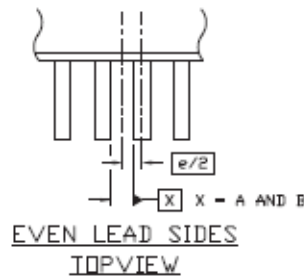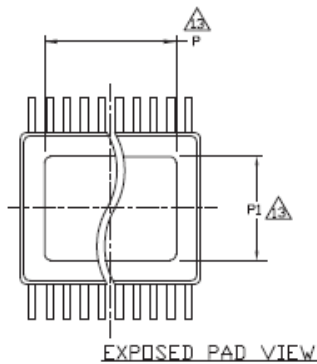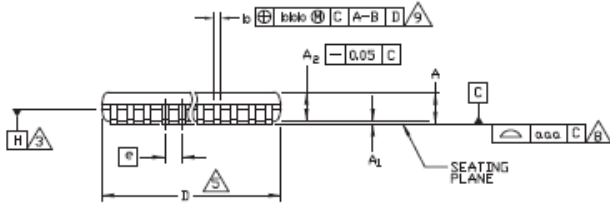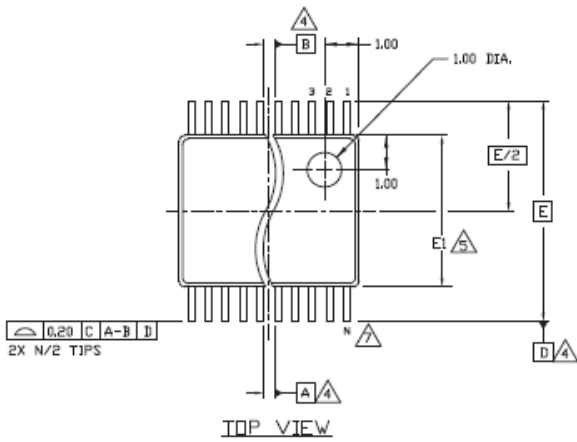
Ordering Information

Table 10. Ordering Information

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
843001CGI-23LF	ICS43001CI23L	"Lead-Free" 24 Lead TSSOP, E-Pad	Tube	-40°C to 85°C
843001CGI-23LFT	ICS43001CI23L	"Lead-Free" 24 Lead TSSOP, E-Pad	Tape & Reel	-40°C to 85°C

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

Revision History Sheet

Rev	Table	Page	Description of Change	Date
A	T10	23	Ordering Information - removed leaded devices. Updated data sheet format.	11/17/15

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